Fault Diagnosis Method of Voltage Sensor in 3-phase AC/DC PWM Converters

Hyung-Seop Kim *, Won-Sang Im *, Jang-Mok Kim *, Dong-Choon Lee ** , and Kyo-Beum Lee ***

Abstract – This paper proposes a fault diagnosis method of the line-to-line voltage sensors in 3-phase AC/DC pulse width modulation (PWM) converters. The line-to-line voltage sensors are an essential device to obtain the information of the grid voltages for controlling the 3-phase AC/DC PWM converters. If the line-to-line voltage sensors are mismeasured by various faults, the voltage sensors can obtain wrong information of the grid voltage. It has an adverse effect on the control of the converter. Therefore, the converter causes the unbalance input AC current and the DC-link voltage ripple in the 3-phase AC/DC PWM converter. Hence, fast fault detection and fault tolerant control are needed. In this paper, the fault diagnosis method is proposed and verified through simulations and experiments.

Keywords: 3-phase PWM converter, Fault diagnosis, voltage sensor

1. Introduction

3-phase AC/DC pulse width modulation (PWM) converters have been increasingly employed in recent years due to their advantages such as a unity power factor and a controllable high-quality dc output voltage [1]. The reliability of the converters has been always important issue due to the costs of unplanned stops [2-3]. Hence the converters require a fault immune characteristic which must have fault detection and tolerant control function [4-7]. Specifically, this paper focused on a line-to-line voltage sensor fault among the various faults. When a fault of the line-to-line voltage sensors occurs, a wrong voltage is measured even if the input grid voltage is healthy. Accordingly, the wrong value becomes the input of the phase locked loop (PLL). Therefore, it is impossible to control the converter normally. Hence the converter causes the unbalanced input AC currents and the DC-link voltage ripple. These may result in malfunctions of the PWM converter or cause secondary faults. Accordingly, there is high possibility of secondary fault in the converter system, the load and the grid [8]. Hence, fast fault detection and fault tolerant control are needed. There are many papers about 3-phase AC/DC converter, but there are few papers

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2. Fault diagnosis method of voltage

2.1 Influence of fault in voltage sensor

Fig. 1 shows the block diagram of the 3-phase AC/DC PWM converter with three voltage sensors. It is essential to measure the voltages of the AC input source for controlling the converter. These voltages can be obtained from the voltage sensors. They become the inputs of PLL as shown in Fig. 2. The grid phase-angle (θ) is acquired through d-q transformation and the PI current controller. Fig. 3 shows the control flow of converter. The voltages (V^d, V^q) and

![Fig. 1. 3-Phase AC/DC PWM converter structure](image)
2.2 Classification of fault in voltage sensor

The various faults of a voltage sensor can be broadly classified into two types. Specifically, the one is a gain-fault and another is a null-fault. The former, a gain-fault, is caused by damages of internal resistors and measurement resistor of the voltage sensor. This gain-fault distorts the amplitude of obtained line-to-line voltage by the voltage sensor. The latter, a null-fault, is caused by the open-circuit inside the voltage sensor. If a null-fault occurs, the measured voltage value is convergent up to ‘0’. This paper describes the diagnosis method about a null-fault and a gain-fault in the converter.

2.3 Fault diagnosis of voltage sensor

Fig.4 shows the flowchart of the proposed fault diagnosis method. In step 1, the line-to-line voltages are measured. In step 2, a sum of voltages is calculated. When voltage sensors have no problem, the sum can be ideally represented as (1).

\[
V_{\text{sum}} = V_{ab} + V_{bc} + V_{ca} = 0 \quad (1)
\]

where,

\[
\begin{align*}
V_{ab} &= V_w \sin(\omega t) \\
V_{bc} &= V_w \sin(\omega t - 120) \\
V_{ca} &= V_w \sin(\omega t + 120)
\end{align*}
\]

Fault diagnosis method

As shown in Fig. 3, accordingly the grid phase-angle which is the output of PLL is estimated incorrectly due to the voltage value obtained by the faulted sensor. The wrong grid phase-angle causes the wrong output of d-q transformation. Therefore, the inputs \( I_d^e, I_q^e \), and \( V_d^e, V_q^e \) of the current controller are distorted, and then distorted output of the current controller generates the six distorted outputs in SVPWM. Thus, the converter becomes in abnormal operation. As a result, the converter has the unbalanced input AC current and the DC-link voltage ripple.
However $V_{sum}$ is difficult to be zero due to the effect of measurement errors such as offset, scale, quantization errors and non-linearity. Hence, in step 3, (2) is used to find out whether or not a fault of the sensors occurs.

$$|V_{ab} + V_{bc} + V_{ca}| < thr.1$$  \hspace{1cm} (2)

If $|V_{sum}|$ is larger than threshold.1 one of three voltage sensor has a problem at least. In step 4, whether or not a null-fault occurs is determined. At the same time, which voltage sensor occurs the null-fault is decided. In this fault case, the value and the variation of voltage are ideally convergent up to 0. However, both are difficult to be exactly convergent up to 0 due to forementioned various measurement errors. A null-fault sensor can be detected through comparing the measured voltages and their variations with threshold 2 as (3).

$$\{\Delta V_{ab} \text{ and } V_{ab}\} < thr.2$$  \hspace{1cm} (3-1)

$$\{\Delta V_{bc} \text{ and } V_{bc}\} < thr.2$$  \hspace{1cm} (3-2)

$$\{\Delta V_{ca} \text{ and } V_{ca}\} < thr.2$$  \hspace{1cm} (3-3)

In step 5, which voltage sensor of three occurs a gain-fault is determined. In this case, the sum of 3-phase voltages is not zero but sine wave. If $V_{sum}$ compares with the faulted voltage, they are inphase or antiphase each other, and the amplitudes of them are different as shown in (4).

$$V_{sum} = V_{ab} + V_{bc} + V_{ca} + Z = Z$$  \hspace{1cm} (4)

$$Z = k \cdot V_{a} \sin(\omega t) \text{ or } k \cdot V_{a} \sin(\omega t + 120) \text{ or } k \cdot V_{a} \sin(\omega t - 120)$$

$$(-1 \leq k \leq 1)$$

Thus, the ratios of two voltage slopes and the ratios of two voltage amplitudes are equal at the same time as (5). Therefore, one of (5) must be true under some fault conditions. After that, which voltage sensor has a gain-fault can be determined by using (5).

$$\frac{V_{sum}}{V_{ab}} = \frac{\Delta V_{sum}}{\Delta V_{ab}}$$  \hspace{1cm} (5)

$$\frac{V_{sum}}{V_{bc}} = \frac{\Delta V_{sum}}{\Delta V_{bc}}$$

$$\frac{V_{sum}}{V_{ca}} = \frac{\Delta V_{sum}}{\Delta V_{ca}}$$

In (5), there can be a measurement error of a sensor as above. Because of that, (5) is hard to be true. If one of sensors occurs a gain-fault, it is detected through the comparison with the absolute value and the threshold 3 by (6). In step 6, the wrong measured value can be replaced with the sum of the rest two sensors as (7).

$$\frac{V_{sum}}{V_{ab}} = \frac{\Delta V_{sum}}{\Delta V_{ab}} < thr.3$$  \hspace{1cm} (6-1)

$$\frac{V_{sum}}{V_{bc}} = \frac{\Delta V_{sum}}{\Delta V_{bc}} < thr.3$$  \hspace{1cm} (6-2)

$$\frac{V_{sum}}{V_{ca}} = \frac{\Delta V_{sum}}{\Delta V_{ca}} < thr.3$$  \hspace{1cm} (6-3)

$$V_{ab} = -(V_{bc} + V_{ca})$$  \hspace{1cm} (7-1)

$$V_{bc} = -(V_{ca} + V_{ab})$$  \hspace{1cm} (7-2)

$$V_{ca} = -(V_{ab} + V_{bc})$$  \hspace{1cm} (7-3)

This method can be implemented quickly within only two sampling time because the variations of voltage requires the present value and the previous one as shown in (8).

$$\Delta V_{sum} = V_{sum}(n) - V_{sum}(n-1)$$

$$\Delta V_{ab} = V_{ab}(n) - V_{ab}(n-1)$$

$$\Delta V_{bc} = V_{bc}(n) - V_{bc}(n-1)$$

$$\Delta V_{ca} = V_{ca}(n) - V_{ca}(n-1)$$  \hspace{1cm} (8)

### 3. Simulation

This section verifies validity of the proposed fault diagnosis method through several simulations. The first example (Ex.1) is that the value of the $V_{ab}$ decreases by

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**Fig. 5.** Three line-to-line voltages (Ex.1)
0.9 times due to a gain-fault of $V_{ab}$ sensor. The results of step 1 are shown in Fig. 5. The sum of line-to-line voltages can be obtained. In step 2, 3, $|V_{sum}|$ is compared with threshold 1 as shown in Fig. 6. Because $|V_{sum}|$ is larger than threshold 1, a occurrence of fault is decided. In step 4, because $|\Delta V_{ab}|$ and $|V_{ab}|$ are upper than threshold 2, a null-fault doesn’t occur. Fig.7(a) shows the results of step 5. Because the value of (6-1) is lower than threshold 3, a gain-fault in the sensor of $V_{ab}$ is regarded as the fault. In the last step 6, the faulted $V_{ab}$ can be replaced with the sum of $V_{bc}$ and $V_{ca}$. For another example(Ex.2), $V_{ca}$ is measured by 1.1 times due to a gain-fault of a sensor as shown in Fig. 8. After a fault, because the results of (6-3) are equal to Fig. 9, a gain-fault of the $V_{ca}$ sensor can be detected. Hence $V_{ca}$ is obtained through the sum of $V_{ab}$ and $V_{bc}$. Therefore, value of fault voltage sensor is replaced with the calculated value by using the two healthy sensors. As a result, the imbalance of the input AC current and the ripple of the output DC voltage disappear.

5. Experiments

Experiments are implemented to verify the algorithm of the proposed fault diagnosis method under the same conditions as simulations. As the first example (Ex.1) of simulations, $V_{ab}$ is measured by 0.9 times when a gain-fault occurs in the sensor of $V_{ab}$. Fig. 10 shows three line-to-line voltages. As shown in Fig. 11, one of three sensors has a fault in step 3. (3) is unsatisfied as shown in Fig. 12.
Thus, when $|abV|$ and $|abV\Delta|$ are higher than threshold $2$, it can be regarded as a gain-fault. The results of step 5 are shown in Fig. 13. Since (a) is true, the sensor of $abV$ is gain-fault. The result of step 6 is shown in Fig. 14. The faulted value of $abV$ is replaced with the calculated value by using the two healthy sensors (7-1). Next example (Ex.2) is the same as the condition of the second simulation, which the amplitude of $caV$ increases by 1.1 times. Fig. 15 shows (Ex.2). As shown in Fig. 16, a gain-fault of $caV$ sensor is confirmed by using (6-3). In the last experiment (Ex.3) as shown in Fig. 17, a sensor of $abV$ is regarded as a null-fault by using (3-1). Fig. 18 shows the
results. The validity of the proposed algorithm is verified through three experiments.

6. Conclusion
This paper analyzes a fault of voltage sensor case and proposes a fault diagnosis method of voltage sensor in 3-phase AC/DC PWM converter. The proposed fault tolerant control method can improve the performance and operate normally when a fault of voltage sensor is detected. This method is easily implemented without any additional H/W by using S/W. The usefulness of the proposed method is verified through simulations and experiments.

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References

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